

### ICT course: Mobile Wireless Communications

#### Lecturers: Dr. Nguyen Minh Huong





# **Course Schedule**

#### • Lectures:

- 1. Introduction
- 2. Characteristics of mobile radio environment:
  - Propagation
  - Fading and mitigations
- 3. Cellular concept
- 4. Channel assignment (optional)
- 5. Modulation techniques
- 6. Multiple Access techniques
- 7. Coding for error detection and correction
- 8. Applications Mobile network Generations:
  - GSM
  - 3G/LTE-4G
  - 5G and future of mobile networks (discussion)
- Exercises

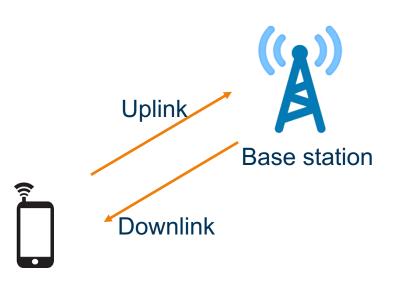
#### • References:

[1]. Mischa Schwartz: Mobile Wireless Communication, CAMBRIDGE UNIVERSITY PRESS, 1st Edition (2005)

[2]. Wireless Communications: Principles and Practice (2nd Edition) by Theodore S. Rappaport

### Lecture 2: Radio environment (wireless channels)

- Propagation phenomena
   Propagation models
   Random Channel Characterization
- □Fading:
  - What is fading? Fading rate?
  - Causes
  - Impacts of fading to transmission
  - Mitigations to fading

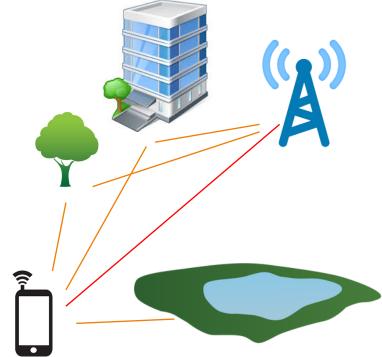


Terminal

# **Propagation Phenomena**

- Electromagnetic signals vary randomly at receivers because of:
  - □ Obstacles → reflection, diffraction, scattering
  - Terminal moves 

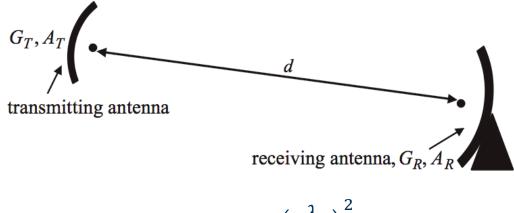
     signal amplitudes fluctuate randomly
  - $\rightarrow$  Fading
- Propagation model captures the variation of power along distances in wireless communication



# **Propagation models**

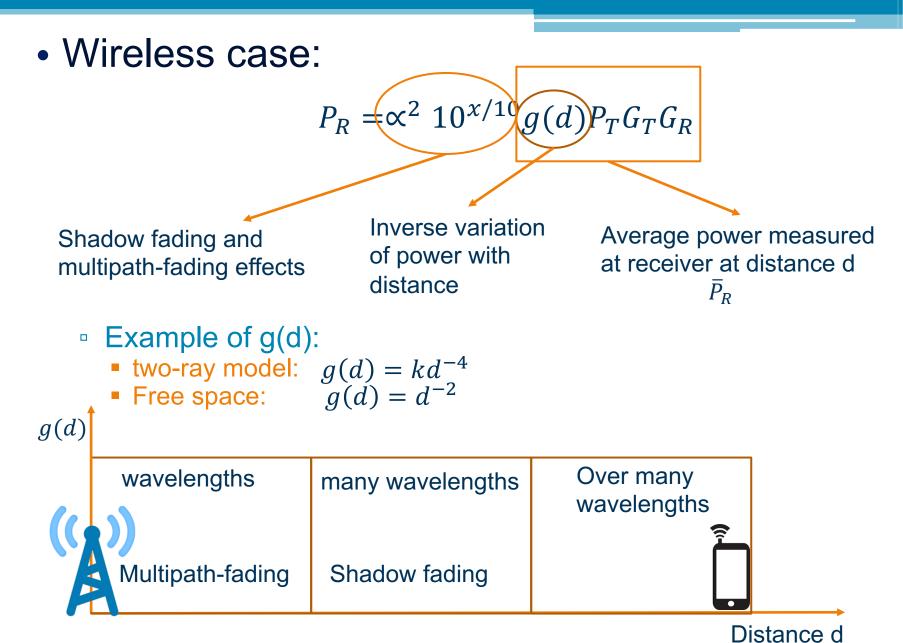
## **Propagation Models**

- Three common effects in wireless communication:
  - Average power varying
  - Long-term variation of average power (Shadow fading)
  - Power variation in wavelength scale (Short-term multipath fading)
- Free space propagation:



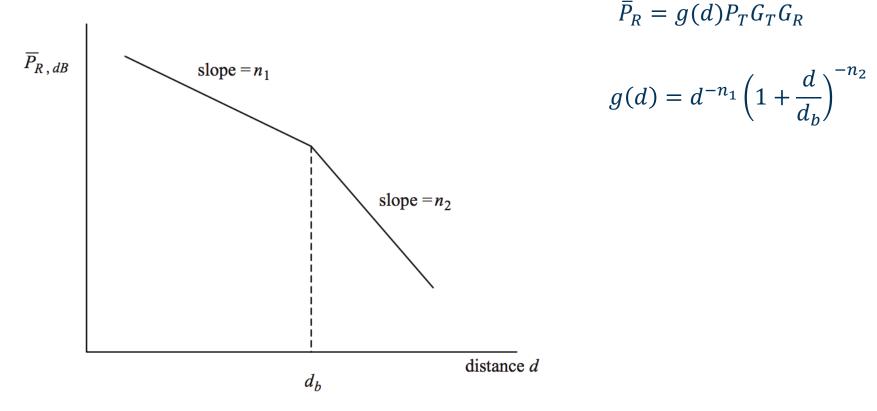
$$P_R = P_T G_T G_R \left(\frac{\lambda}{4\pi d}\right)^2$$

#### **Propagation Models**



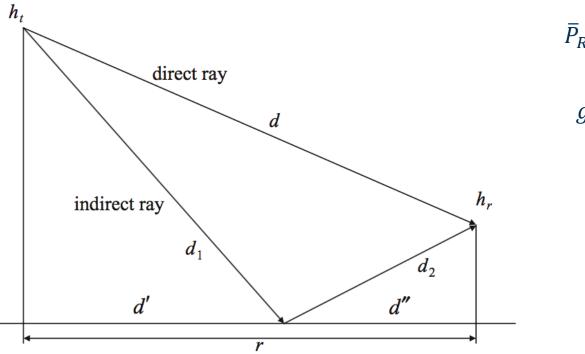
### Propagation models: Average receive power

- Average receive power:  $\overline{P}_R$ 
  - Path loss:
    - Two-slope received signal model: for microcells



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Path loss: two-ray model:



$$\bar{P}_R = g(d) P_T G_T G_R$$

$$g(d) = \frac{(h_t h_r)^2}{d^4}$$

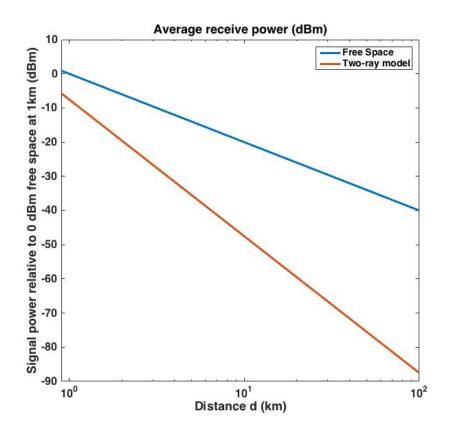


Figure of average power vs distance with different ht

### Propagation models: Shadow (Log-normal) fading

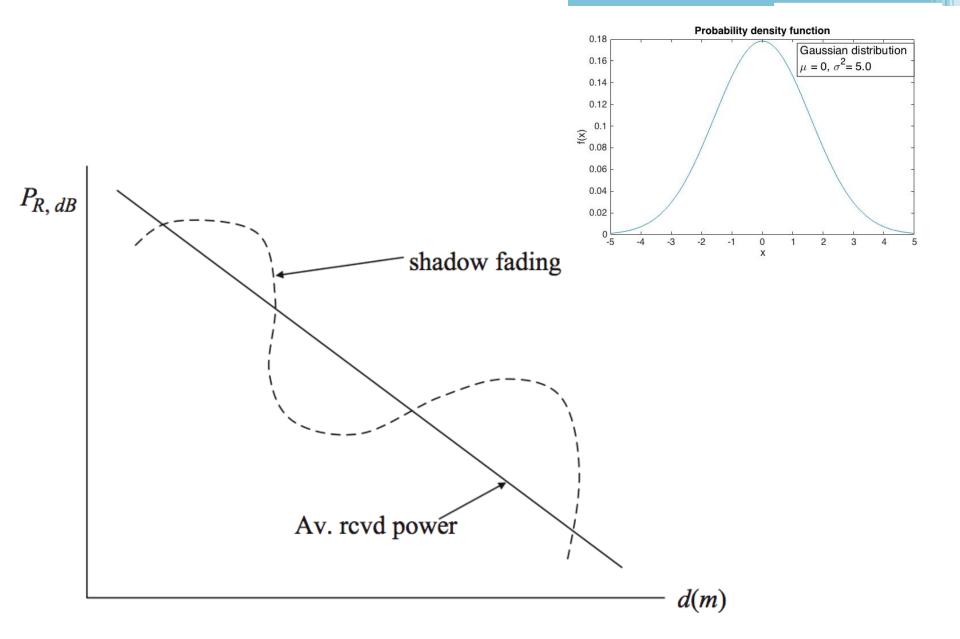
- Shadow-fading (large scale variation):  $10^{x/10}$ 
  - Slowly varying: on the order of meters
  - Over relatively long distance: many wavelengths
  - Average receive power:
  - $\overline{P}_{R,dB} = 10 \log_{10} \overline{P}_R = 10 \log_{10} P_T + 10 \log_{10} g(d) + 10 \log_{10} G_T G_R$

• Receive power:  

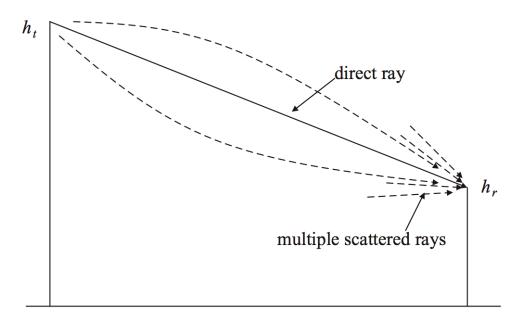
$$P_{R,dB} = 10 \log_{10} P_R = 10 \log_{10} \propto^2 + x + \overline{P}_{R,dB}$$
Local-mean power  $p_{dB}$ 

• Shadow-fading random variable x (dB) is a zeromean Gaussian rv with variance  $\sigma^2$ 

$$f(x) = \frac{e^{-x^2/\sigma^2}}{\sqrt{2\pi\sigma^2}} \qquad f(p_{dB}) = \frac{e^{-(p_{dB} - \bar{P}_{R,dB})^2/\sigma^2}}{\sqrt{2\pi\sigma^2}}$$



#### Propagation models: Rayleigh and Ricean Models



- Over distance of wavelengths
- The direct ray is made up of superposition of scattered L rays

$$S_{R}(t) = \sum_{k=1}^{L} a_{k} \cos[\omega_{c}(t - t_{0} - \tau_{k}) + \theta_{k}]$$
  

$$t_{0} = d/c \qquad \text{Random} \qquad \text{Delay} \qquad \text{Phase} \\ \text{variation} \qquad \text{variation}$$

#### **Propagation Models - Rayleigh model**

$$S_R(t) = a \cdot \cos[\omega_c(t - t_0) + \theta]$$

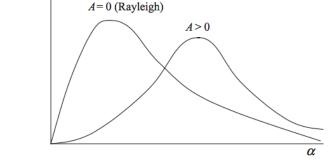
• *a* is Rayleigh distributed (proved) •  $\propto = \sqrt{\frac{c}{2p}}a$  is Rayleigh distributed (proved):  $pdf(\propto) = \frac{\alpha}{\sigma_r^2}e^{-\alpha^2/2\sigma_r^2}$ ,  $\propto \geq 0$  $f_{P_R}(P_R)$  $f(\alpha)$  $\frac{1}{\sigma}e^{-\frac{1}{2}}$  $\sigma$ 0  $P_R$ Exponential distribution Rayleigh distribution • Receive power  $P_R = \frac{ca^2}{2p} = \propto^2 p$  then:  $pdf(P_R) = pdf(\alpha) \cdot \frac{d \alpha}{dP_P} = \frac{1}{n}e^{-P_R/p}$ 

- Ricean model:
  - Distance between two antennas is not too large (e.g., microcellular case)
  - At the receiver: one direct ray and scattered L rays

$$S_R(t) = Acos\omega_c(t - t_0) + \sum_{k=1}^{L} a_k \cos[\omega_c(t - t_0 - \tau_k) + \theta_k]$$
$$S_R(t) = a^* \cos[\omega_c(t - t_0) + \theta]$$

 $f(\alpha)$ 

•  $a^*$  is Ricean distributed (proved)



 Receive power varying about local-mean power

$$f_{P_R}(P_R) = \frac{(1+K)e^{-K}}{p} e^{-\frac{1+K}{p}P_R} I_0\left(\sqrt{\frac{4K(1+K)}{p}P_R}\right)$$

### **Exercise 1.1:**

#### Empirical power drop-off values

City	<i>n</i> <sub>1</sub>	<b>n</b> 2	$d_b(m)$
London	1.7-2.1	2-7	200-300
Melbourne	1.5-2.5	3-5	150
Orlando	1.3	3.5	90

• plot  $P_{R,dB} - P_{T,dB}$  for Orlando as a function of distance *d*, in meters, with 0 < d < 200 m. Assume transmitter and receiver antenna gains are both 1; average power effect is experienced, two-slope model is used.

### **Exercise 1.2:**

The average power received at mobiles 100 m from a base station is 1 mW. Log- normal, shadow, fading is experienced at that distance. The log-normal standard deviation  $\sigma$  is 6dB.

(a) What is the probability that the received power at a mobile at that distance from the base station will exceed 1 mW? Be less than 1 mW?

- (b) An acceptable received signal is 10 mW or higher. What is the probability that a mobile will have an acceptable signal?
- Repeat for  $\sigma = 10 \text{ dB}$ .
- Repeat both cases for an acceptable received signal of 6 mW.

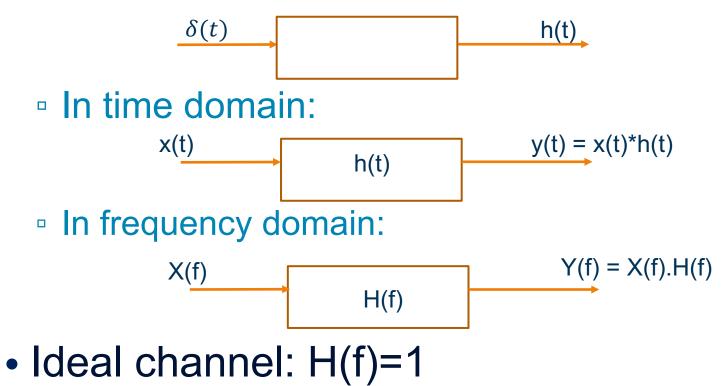
Note: The integral of a Gaussian error function

#### **Random channel characterization**

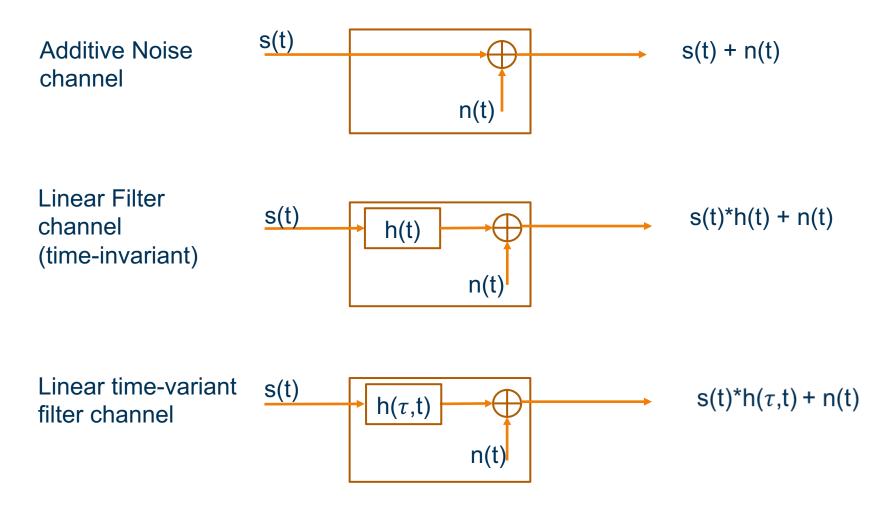
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#### **Random channel characterization**

- System response -Transfer function of a linear system
  - Impulse response







### **Exercise 2.1:**

• Determine output signal of below system:

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$$x(t) = e^{t}[u(t-1) - u(t-3)$$
$$h(t) = \delta(t-1)$$
$$x(t) \qquad h(t) \qquad y(t)=?$$

Determine its transfer function

#### Note: u(t) is a rectangular function



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# Fading

- What is fading?
  - The variation of a transmitted signal at the receiver

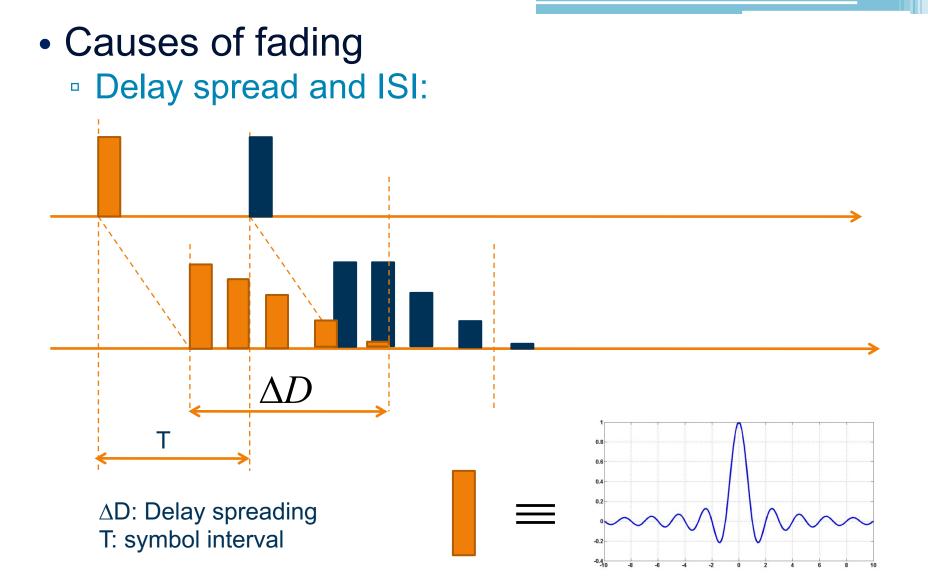
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- Causes of fading
  - Delay spread and Inter-symbol Interference (ISI)Doppler effect

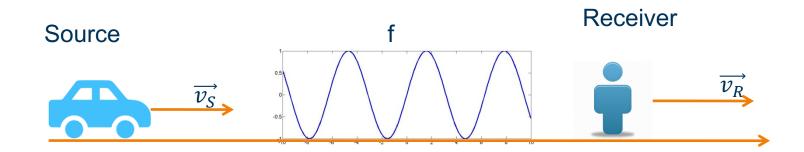
#### • Impacts of fading on transmitted signal:

- Coherence bandwidth and coherence time
- Frequency dispersion: Frequency-selective fading
- Time dispersion: Time-selective fading
- Mitigations to fading

#### Fading: Causes of fading

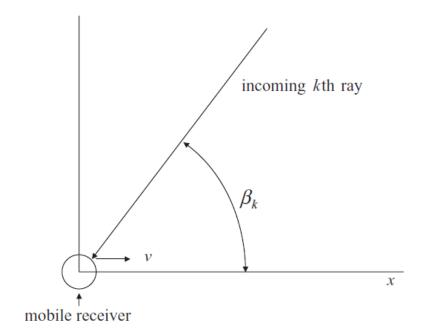


- Doppler effect:
  - General case:



$$f_{R} = f + \frac{v_{R} - v_{S}}{c} f$$
$$\Delta f = \frac{\Delta v}{c} f = \frac{\Delta v}{\lambda}$$

#### Considering terminal mobility



$$f_{Rc} = f_c + f_k$$
$$f_k = \frac{v}{\lambda} \cos \beta_k$$
$$w_k = 2\pi f_k$$

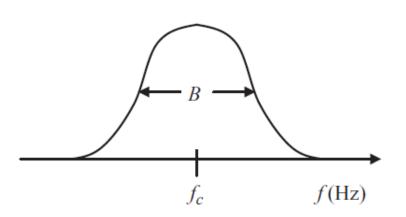
#### Received signal:

$$S_{R}(t) = \sum_{k=1}^{L} a_{k} \cos[w_{c}(t-t_{0}) + \phi_{k} + w_{k}t]$$

#### Fading: Impacts of fading on transmitted signal

#### Fading

- Impacts of fading on transmitted signal
  - How propagation path impacts on the reception of a signal?



(a) Modulated carrier spectrum

(b) Two-carrier model

 $f_2$ 

f(Hz)

 $\Delta f = B$ 

 $f_1$ 

#### Signal bandwidth

- Two carriers represent two frequency components in the spectrum of a real information-bearing signal.
- Frequency separation = bandwidth of the real signal
- The real signal is distorted by the channel if two received signals are not correlated
- I.e., frequency components fade in different ways

# • Received signals $S_{1}(t) = \sum_{k=1}^{L} a_{k} \cos[\omega_{1}(t - t_{1} - \tau_{k}) + \omega_{k}t + \theta_{k}]$ $S_{2}(t) = \sum_{l=1}^{M} a_{l} \cos[\omega_{2}(t - t_{2} - \tau_{l}) + \omega_{l}t + \theta_{l}]$

 $t_1, t_2$ : large-scale delays of two received signals  $\tau_k, \tau_l, \theta_k, \theta_l$ : delays and additional angles of each scattered ray

$$\omega_k = \frac{2\pi(\cos\beta_k)\nu}{\lambda}, \omega_l = \frac{2\pi(\cos\beta_l)\nu}{\lambda}$$

$$\Delta t = t_2 - t_1; \ \Delta f = f_2 - f_1$$

 $f_m = \frac{v}{\lambda}$ : The maximum Doppler frequency

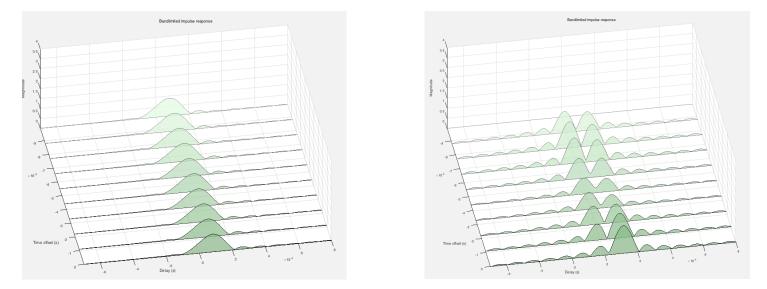
- The coherence depends on bandwidth of the signal and the relative variation in time of the signal to the channel:
  - Correlation function : quantizes the coherence of two random variables (rvs) (How similar?)

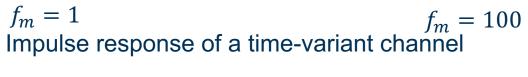
 $p_a = 0$  uncorrelated  $p_a = 1$  Completely correlated

 Using Rayleigh model and exponential incremental delay, correlation function is a function of large-scale time delay and frequency separation:

$$p_a(\Delta \tau, \Delta f)$$

- Coherence bandwidth: is the frequency separation on which  $p_a = 0.5$ 
  - If  $\tau$  has exponential distribution with average value  $\tau_{av}$ and  $\Delta t = 0$ 
    - Coherence bandwidth =  $\frac{1}{2}\pi\tau_{av}$
    - $\tau_{av}$  is average delay spread
- Coherence time:
  - Channel is a time-variant system due to Doppler effect.
  - Coherence time is the time over which the channel impulse response is considered to be unvarying  $\leftrightarrow p_a = 0.5$
  - If  $\tau$  has exponential distribution with average value  $\tau_{av}$
  - Considering the delay spread of one signal:  $\Delta f = 0$ 
    - Coherence time =  $0.18/f_m$

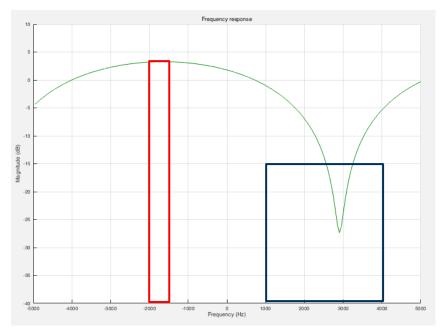




### • Frequency-selective fading:

- Different frequency components experience uncorrelated fading
- Occurs when:

#### Signal Bandwidth > Coherence Bandwidth



#### • Flat fading: when signal bandwidth is narrow

Frequency components experience correlated fading

### • Time-selective fading:

- The channel is time-variant due to Doppler effect
   →At different time, the fading varies
- →One transmitted signal experiences different fading conditions
- Occurs when:

Signal symbol interval > Coherence time

### In conclusion:

Fading channel can distort or cancel the transmitted signal.

 $\rightarrow$  The received signal differs from the original.

→ Receiving errors and/or nothing

# **Exercise 3.1**

### Consider several a delay spread of

- 0.5 msec
- I msec
- 6 msec

Determine whether individual multipath rays are resolvable for the two transmission bandwidths:

- 1.25 MHz used in IS-95 and cdma2000
- 5 MHz used in WCDMA

Note: Relation between bandwidth and symbol interval

B.T = 1 (Fourier analysis)

# **Exercise 3.2**

 Indicate the condition for flat fading for each of the following data rates: 40

- 8 kbps
- 40 kbps
- 100 kbps
- 6 Mbps
- Indicate which, if any, radio environments would result in flat fading for each of these data rates.

# **Fading:** Mitigation to fading

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- Mitigation to fading
  - Channel equalization
  - Diversity reception
  - RAKE time-diversity scheme
  - Others:
    - Interleaving
    - OFDM
    - Coding techniques

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## Mitigation to fading: Equalization technique

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- Equalization technique:
  - Carried out at baseband after demodulating received signal
  - Idea:

$$X(f)$$
 Channel  $Y(f)$   
 $H(f)$ 

- No distortion if: Y(f) = X(f)
- However, Y(f) = X(f). H(f) and  $H(f) \neq 1$
- Solution:

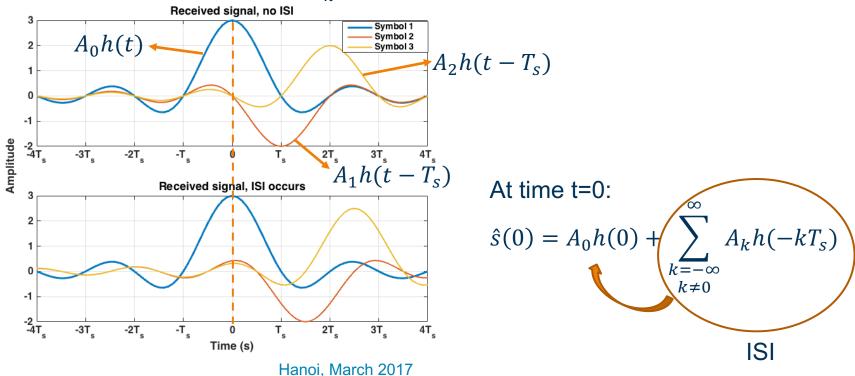
$$X(f)$$
 Channel  $H(f)$   $\frac{1}{H(f)}$   $Y(f)$ 

- Operation:
  - Transmitted signal: Pulse Amplitude Modulation (PAM)

$$s(t) = \sum_{k=-\infty}^{\infty} A_k \delta(t - kT_s)$$

• Received signal after channel:  $\hat{s}(t) = s(t) * h(t)$ 

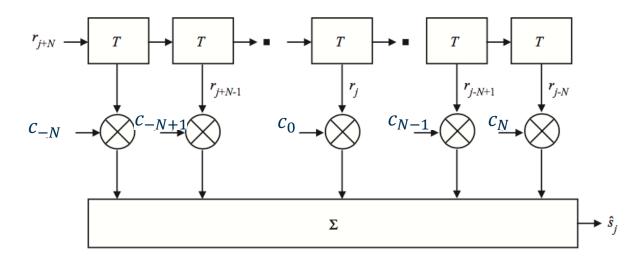
$$\hat{s}(t) = \sum_{k=-\infty}^{\infty} A_k h(t - kT_s)$$



• Canceling the ISI:  $c = (c_{-N}, c_{-N+1}, ..., c_0, c_1, c_2, ..., c_N)$ 

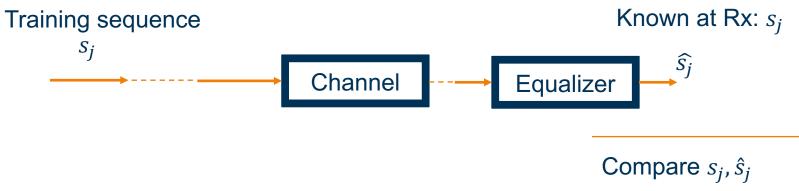
$$A_0 h(0). c_0 = r_j. c_0 = 1$$
$$\sum_{\substack{k=-N\\k\neq 0}}^N A_k h(-kT_s). c_k = \sum_{\substack{k=-N\\k\neq 0}}^N r_{j-k}. c_k = 0$$

• At each symbol interval j:



#### Transversal filter Equalizer

- How to identify c?
  - Training sequence K bits long: known at both Tx and Rx
  - Compare transmitted and estimated sequence of bits:



How to compare?

Min. Mean-Squared Objective Find  $c_k$ ,  $-N \le k \le N$ , such that  $\sum_{j=1}^{K} (s_j - \hat{s}_j)^2$  is minimum

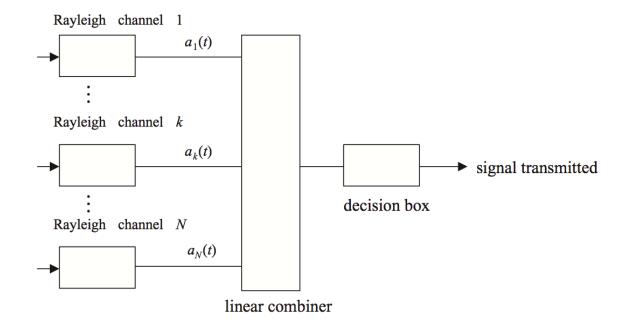
## Mitigation to fading: Diversity reception

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- Diversity reception:
  - Space diversity
  - Frequency diversity
  - Time diversity: RAKE-receiver
  - Angle diversity
  - Polarization diversity

#### General form of diversity reception



- Frequency diversity:
  - OFDM
  - Spread Spectrum
- RAKE receiver:
  - Improving performance of wideband wireless systems
  - combining separately arriving rays of a signal transmitted over a fading channel